



Answer the following questions:

Q1: (a) Describe the output voltage of the circuit shown in Fig.(1). Assuming the diodes to be actual and $V_1 = 10 \sin \omega t$. Sketch one cycle of the output voltage.

(b) Fig.(2) shows a simpler way to draw a transistor circuit. What are collector-emitter voltage and the transistor power dissipation?

Q2: The zener diode in the circuit of figure (3) has a constant reverse breakdown voltage $V_Z = 8.2V$, for $75mA \leq I_Z \leq 1A$, if $R_L = 9 \Omega$, size R_S so that $V_L = V_Z$ is regulated to (maintained at) $8.2V$ while V_S varies by $\pm 10\%$ percent from its nominal value of $12V$.

Q3: (a) Assuming that the diodes in the circuit of Fig.(4) are ideal, find the values of the labeled voltage, V , and current, I .

(b) What is the output voltage in Fig.(5). Let β of the two transistors are very high.

Q4: The transistor in the circuit shown in fig.(6) is biased to operate in the active mode. Assuming that β is very large, find the collector bias current I_C . Replace the transistor with small-signal equivalent circuit T-model, find the values of the voltage gains of (V_{o1}/v_i) and (V_{o2}/v_i) .

Q5: For the amplifier shown in Fig.(7), let $V_{CC} = 12V$, $R_1 = 22 k\Omega$, $R_2 = 6.8 k\Omega$, $R_E = 560 \Omega$, and $R_C = 1 k\Omega$. The transistor has $\beta = 100$. Calculate the dc bias current I_E . If the amplifier operates between a source for which $R_{sig} = 600 \Omega$ and a load of $2 k\Omega$, replace the transistor with its hybrid- π model, and find the values of R_{in} , R_o , and the voltage gain v_o/v_{sig} .

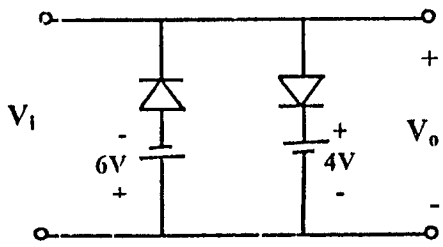


Fig. (1)

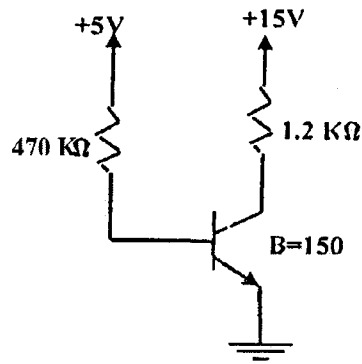


Fig.(2)

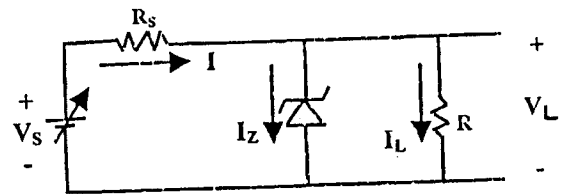


Fig.(3)

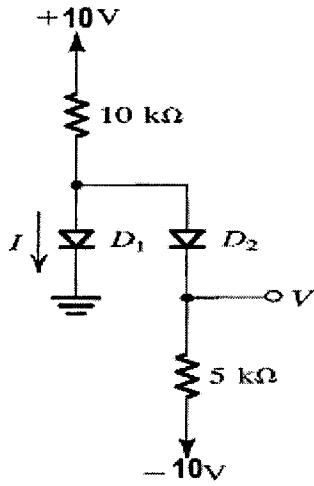


Fig.(4)

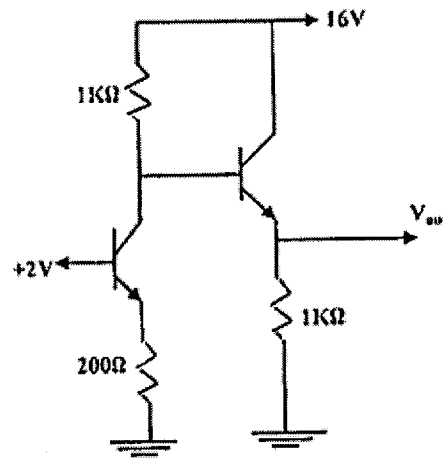


Fig.(5)

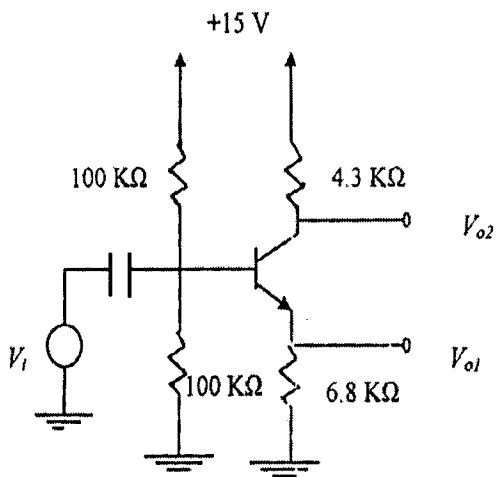


Fig.(6)

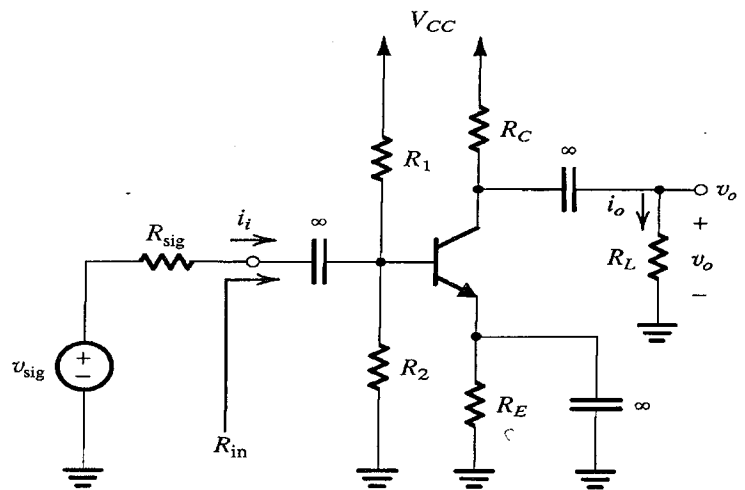


Fig.(7)

BEST WISHES
Hossam Labib

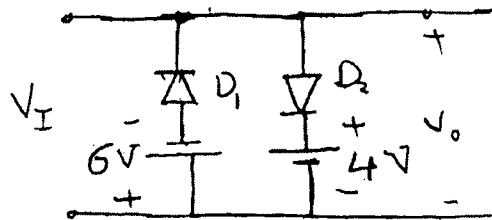
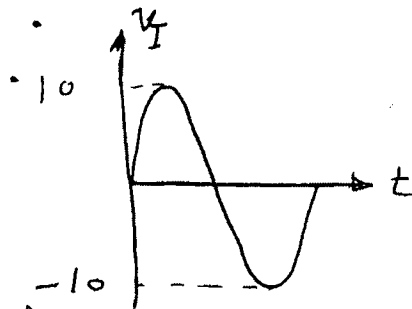
Q1 (a) (6 points)

Describe the o/p voltage of the circuit shown.

Assuming the diode to be actual and $V_I = 10 \sin \omega t$.

Sketch one cycle of the o/p.

Solution

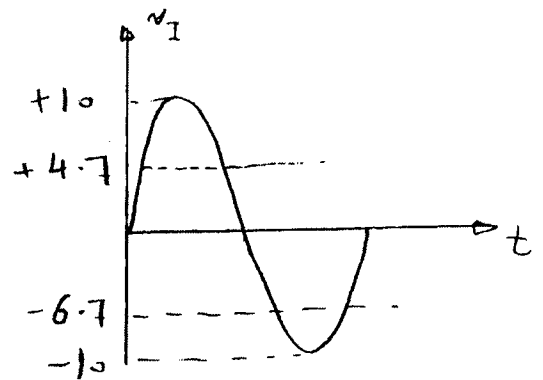
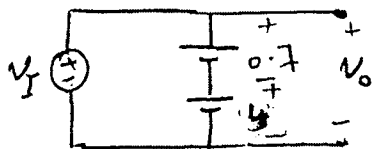


For +ve half cycle

- For $V_i < 4.7$ V
 D_1 and D_2 are off
 $\therefore V_o = V_i$

- For $V_i > 4.7$ V (2)

- D_1 off & D_2 on
 $\therefore V_o = 4.7$ V

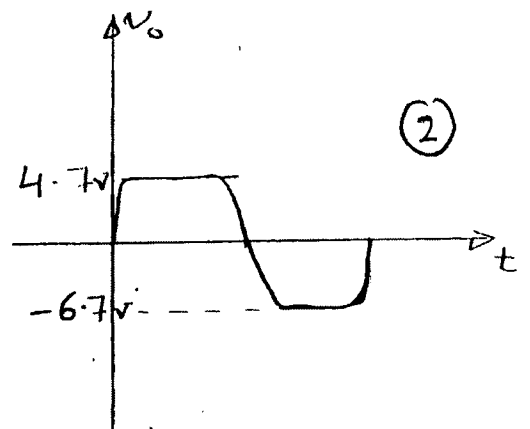
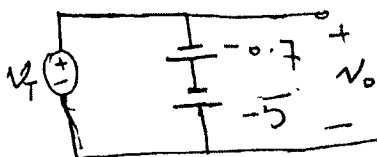


For -ve half cycle

- For $V_i > -6.7$
 D_1 off and D_2 off
 $\therefore V_o = V_i$

- For $V_i < -6.7$ V (2)

- D_1 on & D_2 off
 $\therefore V_o = -6.7$



Q₁(b) (1 point)

Fig. (2) shows a simpler way to draw a Transistor circuit. What are collector emitter voltage and the Transistor power dissipation?

Solution

Let Transistor in Active Region

Loop (I)

$$5 = 470K I_B + V_{BE}$$

$$I_B = \frac{5 - 0.7}{470K} = 9.15 \mu A$$

$$\therefore I_C = \beta I_B = 150 * 9.15 \mu A = 1.3725 \text{ mA}$$

$$V_{CE} = V_{CC} - I_C R_C = 15 - 1.3725 * 1.2 = 13.353 \text{ V}$$

$$V_E = 0 \text{ V} \quad \therefore V_{BE} = V_B - V_E$$

$$\therefore V_B = V_{BE} = 0.7 \text{ V} \quad ; \quad V_C = V_{CE} = 13.353 \text{ V}$$

$\therefore V_B > V_E \Rightarrow$ BE \downarrow Forward bias

$\therefore V_B < V_C \Rightarrow$ BC \downarrow Reverse bias

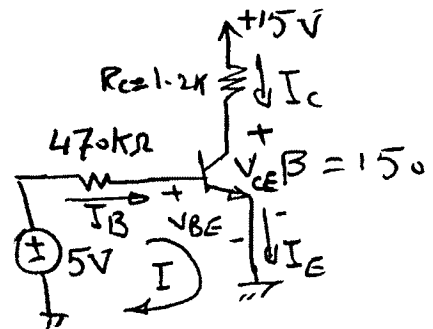
\therefore Transistor in Active Region

$$\therefore V_{CE} = 13.353 \text{ V} \quad ; \quad I_C = 1.3725 \text{ mA}$$

$$P_{diss} = I_C \cdot V_{CE}$$

$$= 1.3725 * 10^{-3} * 13.353 = 0.0183 \text{ W}$$

$$= 18.3 \text{ mW}$$

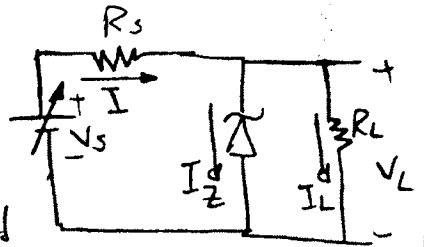


Q2: (1 points)

The Zener diode shown has a constant reverse breakdown voltage $V_Z = 8.2 \text{ V}$

for $75 \text{ mA} \leq I_Z \leq 1 \text{ A}$ & $R_L = 9 \Omega$

size R_S so that $V_L = V_Z$ is regulated to (maintained at) 8.2 V while V_S varies by $\pm 10\%$ percent from its nominal value of 12 V



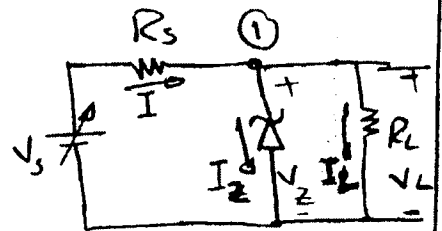
Solution

at node ①

$$I = I_Z + I_L \rightarrow \textcircled{1}$$

$$I_L = \frac{V_L}{R_L} = \frac{V_Z}{R_L} = \frac{8.2}{9} = 0.911 \text{ A}$$

$$\therefore R_S = \frac{V_S - V_Z}{I} = \frac{V_S - V_Z}{I_Z + I_L} \rightarrow \textcircled{2}$$



We use ② to size R_S for max. Zener current I_Z at the largest value of V_S ; $\therefore 75 \text{ mA} \leq I_Z \leq 1 \text{ A}$

$$\text{i.e. } I_Z = 1 \text{ A} ; \quad V_S = V_S + 10\% V_S = 1.1 V_S = 1.1 \times 12 = 13.2 \text{ V}$$

$$\therefore R_S = \frac{13.2 - 8.2}{1 + 0.911} = 2.62 \Omega$$

- We check to see if $I_Z \geq 75 \text{ mA}$ at the lowest value of V_S ; i.e. $V_S = V_S - 10\% V_S = 0.9 V_S = 0.9 \times 12 = 10.8 \text{ V}$.

$$I_Z = I - I_L = \frac{V_S - V_Z}{R_S} - I_L = \frac{10.8 - 8.2}{2.62} - 0.911 = 0.9924 - 0.911 = \underline{81.4 \text{ mA}} > 75 \text{ mA}$$

since $I_Z > 75 \text{ mA}$; $V_Z = 8.2 \text{ V}$ then regulation is occurred (preserved)

Q3(a)

Assuming that the diodes in the circuit of Fig.(2) are ideal, Find the values of the Labeled voltage, V , and current, I .

Solution

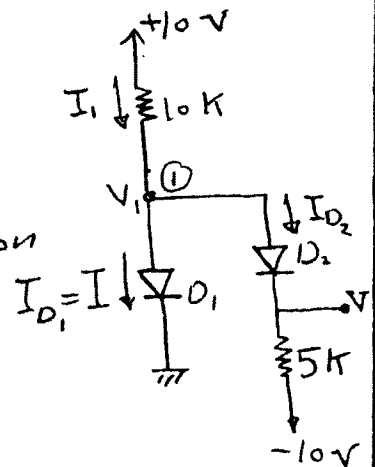
let $I = I_{D_1}$

let D_1 and D_2 are ON

For D_1 ON $\Rightarrow V_1 = 0 \Rightarrow V = 0$ For D_2 ON

$$\therefore I_1 = \frac{10 - V_1}{10K} = \frac{10 - 0}{10K} = 1mA$$

$$I_{D_2} = \frac{V_1 - (-10)}{5K} = \frac{0 + 10}{5K} = 2mA$$



From node ① $\Rightarrow I_1 = I_{D_1} + I_{D_2}$

$$I_{D_1} = I_1 - I_{D_2} = 1mA - 2mA = -1mA$$

$\therefore I_{D_1}$ -ve \therefore NOT TRUE

i.e. the assumption NOT correct

Let D_1 OFF and D_2 ON

- For D_1 OFF $\Rightarrow I_{D_1} = 0$; $I_1 = I_{D_2}$

- For D_2 ON $\Rightarrow I_1 = \frac{10 - (-10)}{10K + 5K} = \frac{20}{15K} = 1.333 mA$

$$V = I_{D_2} * 5K - 10 = 1.333 * 5 - 10 = -3.335 V$$

For $V = -3.335 V$; $\therefore V_1 = V = V_{p1} = -3.335$; $V_{n1} = 0$

i.e. $V_{p1} < V_{n1} \Rightarrow D_1$ OFF

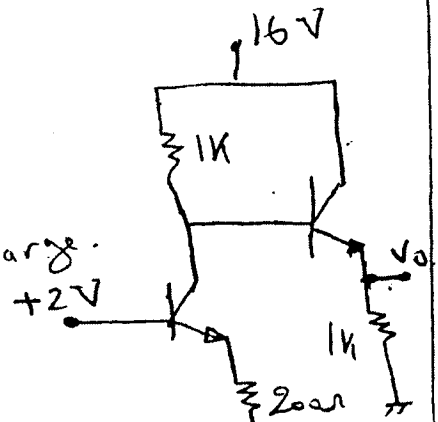
\therefore The assumption is True (D_1 OFF & D_2 ON)

$$\therefore V = -3.335 V \text{ and } I = I_{D_1} = 0 A$$

Q3(b): (points)

what is the o/p voltage.

- Let β of the 2-transistors is very large.



Solution

$\therefore \beta$ is very high

$\therefore I_{B1} = I_{B2} \approx 0$

Loop (I) let Q_1 and Q_2 in Active region

$$-2 + V_{BE1} + I_{E1} R_{E1} = 0$$

$$I_{E1} = \frac{2 - 0.7}{200} = 6.5 \text{ mA}$$

$$\therefore I_{C1} = I_{E1}$$

$$\therefore \text{at node } \textcircled{1} \Rightarrow I_1 = I_{C1} + I_{B2} \quad ; \quad I_{B2} = 0$$

$$\therefore I_1 = I_{C1} = 6.5 \text{ mA}$$

$$V_{C1} = 16 - I_1 \times 1k = 16 - 6.5 \times 1 = 9.5 \text{ V} = V_{B2}$$

$$V_{E1} = I_{E1} R_{E1} = 6.5 \times 0.2 = 1.3 \text{ V}$$

Loop (II)

$$-V_{C1} + V_{BE2} + I_{E2} R_{E2} = 0$$

$$I_{E2} = \frac{9.5 - 0.7}{1k} = 8.8 \text{ mA}$$

$$\therefore V_0 = I_{E2} R_{E2} = 8.8 \times 1 = 8.8 \text{ V} = V_{E2}$$

$$V_{C2} = 16 \text{ V}$$

for Q_1 $\therefore V_{B1} > V_{E1} \Rightarrow B-E \rightarrow$ Forward

$V_{B1} < V_{C1} \Rightarrow B-C \rightarrow$ Reverse

$\therefore Q_1$ in Active region

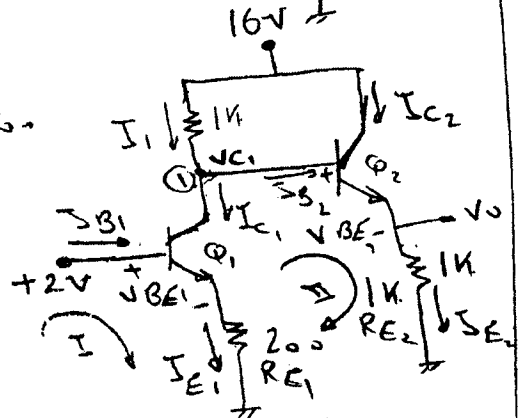
for Q_2 $\therefore V_{B2} > V_{E2} \Rightarrow B-E \rightarrow$ Forward.

$V_{B2} < V_{C2} \Rightarrow B-C \rightarrow$ Reverse

$\therefore Q_2$ in Active region.

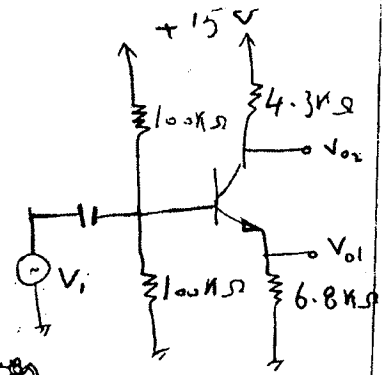
\therefore Assumption is True.

$$\therefore V_0 = V_{E2} = 8.8 \text{ V}$$



Q4: (5 POINTS)

The transistor in the circuit shown is biased to operate in the active mode. Assuming that β is very large, find the collector bias current. Replace the transistor with small-signal equivalent circuit T-Model, find the values of the voltage gains of V_{o1} , V_{o2} .



DC analysis

Reduce AC sources i.e. $V_i = s.c$
and ALL capacitor are o.c

$$V_{Th} = 15 \frac{100k}{100k+100k} = 7.5 \text{ V}$$

$$R_{Th} = 100k \parallel 100k = 50k\Omega$$

β is very large $\Rightarrow I_B = 0$
 $\alpha = 1$ & $I_C = I_E$

Loop (I)

$$-V_{Th} + I_B R_{Th} + V_{BE} + 6.8k I_E = 0$$

$$I_E = \frac{V_{Th} - V_{BE}}{6.8k} = \frac{7.5 - 0.7}{6.8k} = 1 \text{ mA}$$

$$\therefore I_C = 1 \text{ mA}$$

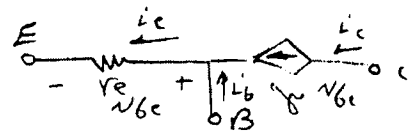
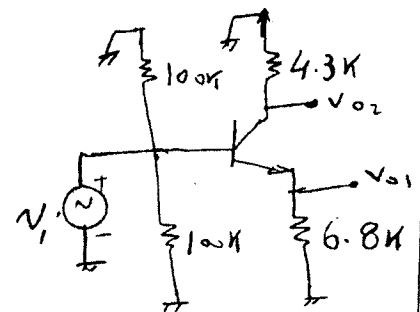
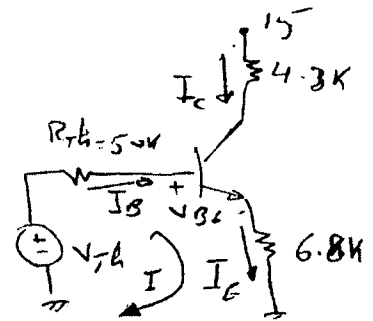
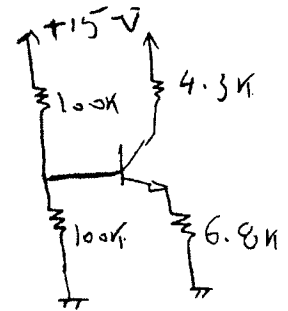
AC analysis

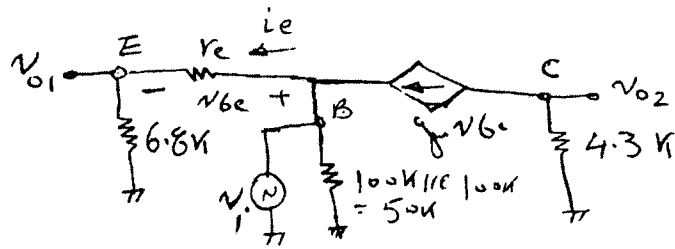
Reduce DC sources $V_{CC} = s.c$
ALL capacitor are s.c

The parameters of T-Model are

$$g_m = \frac{I_C}{V_T} = \frac{1 \text{ mA}}{25 \text{ mV}} = 40 \text{ mA/V}$$

$$r_e = \frac{V_T}{I_E} = \frac{25 \text{ mV}}{1 \text{ mA}} = 25 \Omega$$





$$v_{o2} = -\beta v_{be} \times 4.3K$$

$$v_{be} = v_i \frac{r_e}{r_e + 6.8K}$$

$$v_{o2} = -\beta v_i \frac{r_e \times 4.3K}{r_e + 6.8K}$$

$$\frac{v_{o2}}{v_i} = -\beta \frac{r_e \times 4.3K}{r_e + 6.8K} = -40 \times 10^3 \frac{25 \times 4300}{25 + 6800} = -0.63 \text{ V/V}$$

$$\# v_{o1} = v_i \frac{6.8K}{6.8K + r_e}$$

$$\frac{v_{o1}}{v_i} = \frac{6.8K}{6.8K + 25} = 0.996 \text{ V/V}$$

(Q5)

For the amplifier shown in Fig.(3), let $V_{CC} = 12V$, $R_1 = 22\text{ k}\Omega$, $R_2 = 6.8\text{ k}\Omega$, $R_E = 560\ \Omega$, and $R_C = 1\text{ k}\Omega$. The transistor has $\beta = 100$. Calculate the dc bias current I_E . If the amplifier operates between a source for which $R_{sig} = 600\ \Omega$ and a load of $2\text{ k}\Omega$, replace the transistor with its hybrid- π model, and find the values of R_{in} , R_o , and the voltage gain

v_o/v_{sig} .

Solution

Dc Analysis

- ALL capacitor are o-c
- Reduce AC sources

$$R_{Th} = R_1 \parallel R_2 = 22\text{K} \parallel 6.8\text{K}$$

$$R_{Th} = 5.19\text{K}\ \Omega$$

$$V_{Th} = V_{CC} \frac{R_2}{R_1 + R_2}$$

$$V_{Th} = 12 \frac{6.8\text{K}}{22\text{K} + 6.8\text{K}} = 2.833\text{ V}$$

Loop (I)

$$-V_{Th} + I_B R_{Th} + V_{BE} + I_E R_E = 0$$

$$\therefore I_E = (1 + \beta) I_B$$

$$\frac{I_E}{1 + \beta} R_{Th} + I_E R_E = V_{Th} - V_{BE}$$

$$I_E = \frac{V_{Th} - V_{BE}}{R_E + \frac{R_{Th}}{1 + \beta}} = \frac{2.833 - 0.7}{560 + \frac{5.19\text{K}}{101}} = 3.489\text{ mA} \quad \text{--- (1)}$$

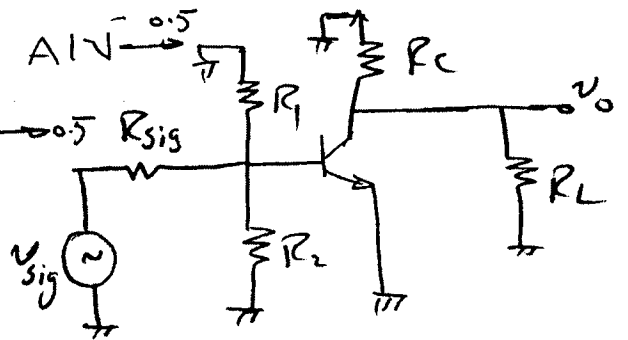
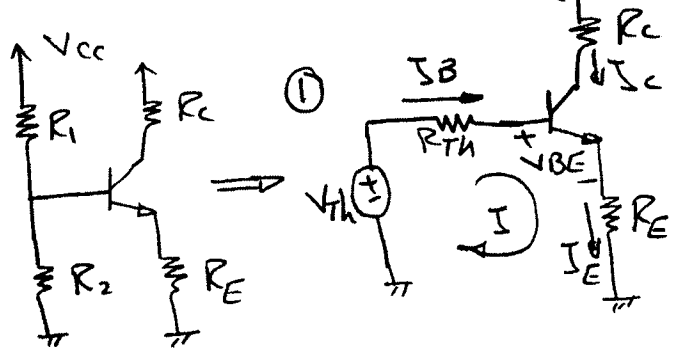
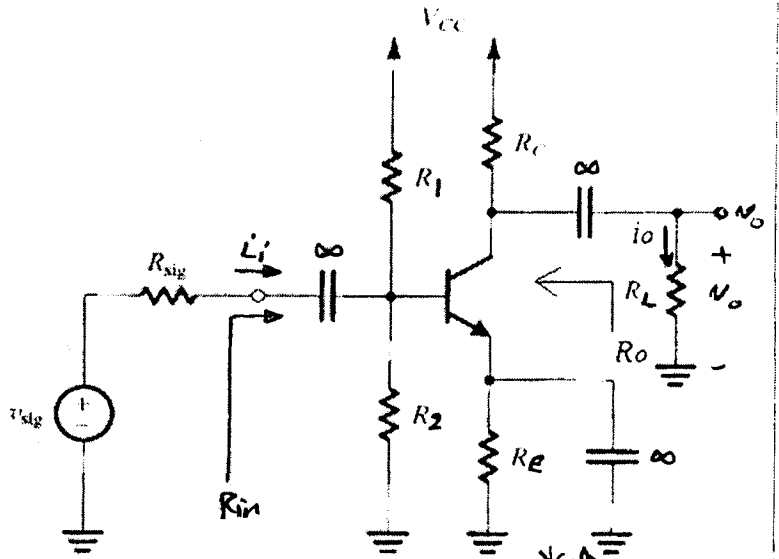
$$I_C = \alpha I_E = \frac{\beta}{1 + \beta} I_E = \frac{100}{101} \times 3.489\text{ mA} = 3.454\text{ mA}$$

$$g_m = \frac{I_C}{V_T} = \frac{3.454\text{ mA}}{25\text{ mV}} = 0.138\text{ A/V} \quad \text{--- (2)}$$

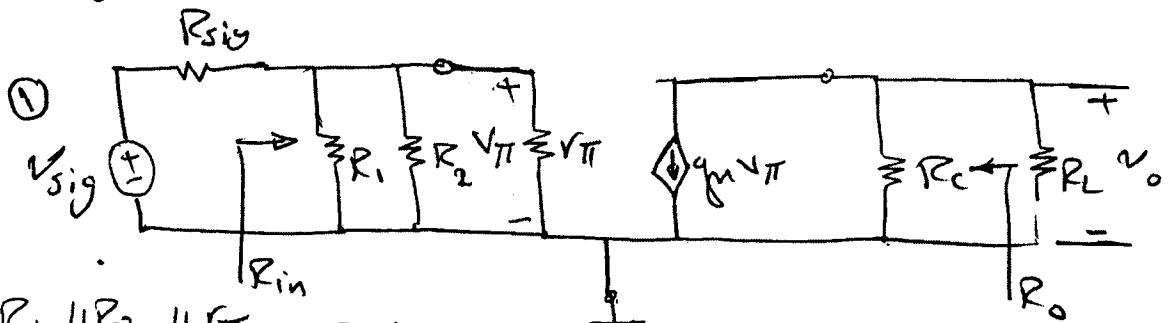
$$r_{\pi} = \frac{\beta}{g_m} = \frac{100}{0.138} = 724.64\ \Omega$$

AC Analysis

- ALL capacitor are s.c
- Reduce DC sources



Using π -model



$$R_{in} = R_1 \parallel R_2 \parallel R_{\pi} = 22K \parallel 6.8K \parallel 724.61$$

$$= 635.93 \approx 636 \Omega \quad \longrightarrow \textcircled{1}$$

$$R_{o|_{v_{sig}=0}} = R_c = 1 K\Omega. \quad \longrightarrow \textcircled{1}$$

To find $\frac{v_o}{v_{sig}}$:-

$$v_o = -g_m v_{\pi} (R_c \parallel R_L)$$

$$v_{\pi} = v_{sig} \frac{R_{in}}{R_{in} + R_{sig}}$$

$$v_o = -g_m v_{sig} \frac{R_{in}}{R_{in} + R_{sig}} (R_c \parallel R_L) \quad \longrightarrow \textcircled{2}$$

$$\frac{v_o}{v_{sig}} = -g_m \frac{R_{in}}{R_{in} + R_{sig}} (R_c \parallel R_L)$$

$$= -0.138 \frac{636}{636 + 600} (1K \parallel 2K)$$

$$= -0.138 * 0.515 * 666.667 = -47.34 \text{ mV} \quad \longrightarrow \textcircled{1}$$