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Answer

Q1 a- Define:  $\omega_n$ ,  $\omega_d$ ,  $\omega_r$ ,  $\omega_B$ ,  $\omega_c$ ,  $\omega_g$ ,  $\omega_p$ ,  $M_r$ ,  $\eta$ ,  $G_m$ ,  $\gamma_m$ ? (15 marks)

-**Natural frequency  $\omega_n$  rad/sec:** it is the natural frequency depends on the natural of the system parameters.

- **Under damped natural frequency  $\omega_d$  rad/sec:** it is the under damped natural frequency depends on the damping coefficient  $\eta$  as it is less than one  $\eta < 1$ .

-**Resonant frequency  $\omega_r$  rad/sec:** it is the frequency at which the peak value of the output frequency response for a second order is equal to  $\omega_r = \omega_n \sqrt{1 - 2\zeta^2}$

$$\omega_r = \omega_n \sqrt{1 - 2\zeta^2}, \quad \text{for } 0 \leq \zeta \leq 0.707$$

As  $\zeta$  approaches zero,  $M_r$  approaches infinity

$0 < \zeta \leq 0.707$ , the resonant frequency  $\omega_r$  is less than the damped natural frequency

-**Cut off frequency  $\omega_B$  rad/sec:** it is the frequency at which the magnitude of the output frequency response is equal to  $(= \frac{1}{\sqrt{2}})$  of the low frequency .

-**Corner frequency  $\omega_c$  rad/sec:** it is the frequency at which the magnitude of the output frequency response is changed sharply. It may be  $(0, 1, 1/T, \omega_n)$

-**Gain crossover frequency  $\omega_g$ :** it is the frequency at which the magnitude of the output frequency response is equal to one or zero decibel.

$$|G(j\omega_g)H(j\omega_g)| = 1 \quad \text{or} \quad |G(j\omega_g)H(j\omega_g)| = 0db$$

-**Phase crossover frequency  $\omega_p$ :** it is the frequency at which the phase of the output frequency response is equal to  $(-180)$  degrees.

$$\text{Imag. } [G(j\omega_p)H(j\omega_p)] = 0 \quad \text{or} \quad \angle G(j\omega_p)H(j\omega_p) = -180\text{deg.}$$

**-Maximum resonant magnitude  $M_r$ :** it is the peak value of the output frequency

response for a second order system  $M_r = \frac{1}{2\zeta\sqrt{1-\zeta^2}}$

$$\frac{C(s)}{R(s)} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2} \quad M_r = |G(j\omega)|_{\max} = |G(j\omega_r)| = \frac{1}{2\zeta\sqrt{1-\zeta^2}}$$

**-damping coefficient  $\eta$**  it depends on the natural of the system parameters. For second order system

$$\frac{C(s)}{R(s)} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

Values of $\eta$	System stability	Step-response
$0 > \eta$	System is unstable	undefined
$\eta = 0$	System is critically stable	oscillatory
$0 < \eta < 1$	System is stable	Under-damped
$0 < \eta = 1$	System is stable	Critically damped
$0 < \eta > 1$	System is stable	Over damped

**-Gain margin  $G_m$ :** it is reciprocal of the magnitude of the output frequency response at the **Phase crossover frequency  $\omega_p$**

$$G_m = 1/[\text{Real of } G(j\omega_p)H(j\omega_p)] = 1/|G(j\omega_p)H(j\omega_p)| = K_c/K$$

$$G_M = 20 \log G_m \text{ db}$$

**-Phase margin  $\gamma_m$ :** it is the angle of the output frequency response at the **gain crossover** frequency plus 180 degrees.

$$\gamma_m = \angle G(j\omega_g)H(j\omega_g) + 180 \text{ deg.}$$

b- Consider a control system shown in Fig.1 if  $G(S) = 4/[S(S+2)]$ ,  $H(S) = 1$

i-Find the **frequency response** as  $r(t) = 5\sin\omega t$  ?      ii-Calculate  $M_r$ ,  $\omega_r$ ?

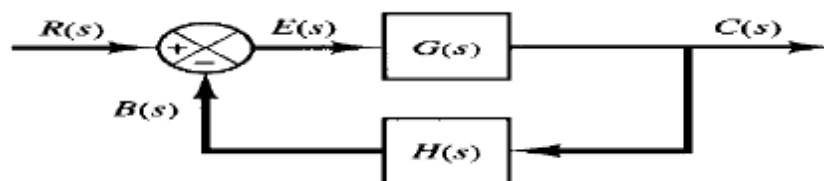


Fig.1

Frequency Response: it means the steady state output of a 1-linear 2-time-invariant 3- stable control system to a sinusoidal input and it is a sinusoidal with phase shift positive or negative and does not depend on the initial conditions.

**b-Steps to find frequency Response:**

1- the closed loop transfer function =  $T(s)=C(S)/R(S) =$

$$C(S) / R(S) = \frac{G(s)}{1+G(S)H(S)} = \frac{\omega_n^2}{s^2+2\eta\omega_n s+\omega_n^2} = \frac{4}{s^2+2s+4}, \quad \omega_n = \frac{2\text{rad}}{\text{sec}} \quad \zeta = 0.5$$

2-the closed loop frequency transfer function =

$$T(j\omega)=C(j\omega)/R(j\omega) = \frac{4}{(j\omega)^2+2(j\omega)+4} = M \angle \Phi = \text{Re}+j \text{ imag}$$

$$M = \frac{4}{\sqrt{(4-\omega^2)^2+4\omega^2}}, \quad \Phi = \tan^{-1}[2\omega / (4 - \omega^2)]$$

3-As the input =  $r(t) = 5\sin\omega t$  then

$$\begin{aligned} \text{the response} = C(t) &= 5M\sin(\omega t + \Phi) \\ &= \frac{20}{\sqrt{(4-\omega^2)^2+4\omega^2}} \sin[\omega t + \tan^{-1}[2\omega / (4 - \omega^2)]] \end{aligned}$$

$$M_r = \frac{1}{2\zeta \sqrt{1 - \zeta^2}} = \frac{1}{2(0.5) \sqrt{1 - (0.5)^2}} = 1.155$$

$$\omega_r = \omega_n \sqrt{1 - 2\zeta^2} = 2\sqrt{1 - 2(0.5)^2} = 1.414 \text{ rad/sec.}$$

Q2

(15 marks)

Consider a control system shown in Fig.1 if  $G(S) = K/[(S+3)(S+2)]$ ,  $H(S) = 1/(S+1)$

- a- Sketch the **complete root locus** for positive values of **K**?
- b- Find **K** that makes the complex closed loop poles have a damping ratio =**0.6** and **find the closed loop poles using the plot**?
- c- Find **K** that makes the complex closed loop poles have a damping ratio =**0.6** and **find the closed loop poles analytically**?
- d- Write short MATLAB program to solve **a** and solve **b**?

Root locus:

1-the root locus is symmetrical about the real axis in the S-plane

2-the open loop TF= $G(s)H(s) = \frac{K}{(S+3)(S+2)(S+1)} = \frac{K}{S^3+6S^2+11s+6}$

3-the root locus starts at the pole and ends at the zero or infinity

4-number of root loci=  $n$ =number of poles of the open loop TF =3 at [-1,-2,-3]

5-number of zeros=  $m=0$

6-number of asymptotes =  $n-m=3-0=3$

8-center of gravity =  $A = \frac{\sum poles - \sum zeros}{n-m} = \frac{-1-2-3}{3} = -2$  point of intersection of asymptotes with real axis=

9-angles of asymptotes are =  $\Theta = \frac{\pm 180(2R+1)}{n-m} = \pm 60, \pm 180$

10- Points of crossing the imaginary axis as Routh test

Charct.equa= $1+G(S)H(S)=0 = S^3+6S^2+11s+6+K$

$S^3$	1	11	$6+K \geq 0, [66-6-K]/6 \geq 0$ then - $6 \leq K \leq 60, K_c = 60$ $6S^2+6+60=0, S=j \omega = \sqrt{11}$ rad/sec
$S^2$	6	$6+K$	
$S$	$[66-6-K]/6$		
$S^0$	$6+K$		

11- break points (break away or break in) at

$$-\frac{dK}{dS} = 0 = \frac{d}{dS} \left[ \frac{1}{G(S)H(S)} \right] = \frac{d}{dS} [S^3 + 6S^2 + 11s + 6] = 3S^2 + 12s + 11 = 0$$

$S=-2.6$  refused,  $S=-1.4$  is a break- away point

12-break angles at  $[\pm 180(2R+1)/r]$  where  $r$ =number of branches (poles for break away or zeros for break in)  $R=0,1,-----$  break angles at  $[\pm 180]/2 = \pm 90$

13-there is no angle of departure (complex poles)

14- there is no angle of arrival (complex zeros)

15-sketch the root loci as

16- the damping factor or coefficient  $\zeta$  is straight line with slope  $\Theta = \cos^{-1} \zeta$

with respect to the negative real axis in the S-plane.  $\Theta = \cos^{-1} 0.6 = 53.13\text{deg.}$  at the test point (intersection point)  $S_d = -1 \pm j1.4$

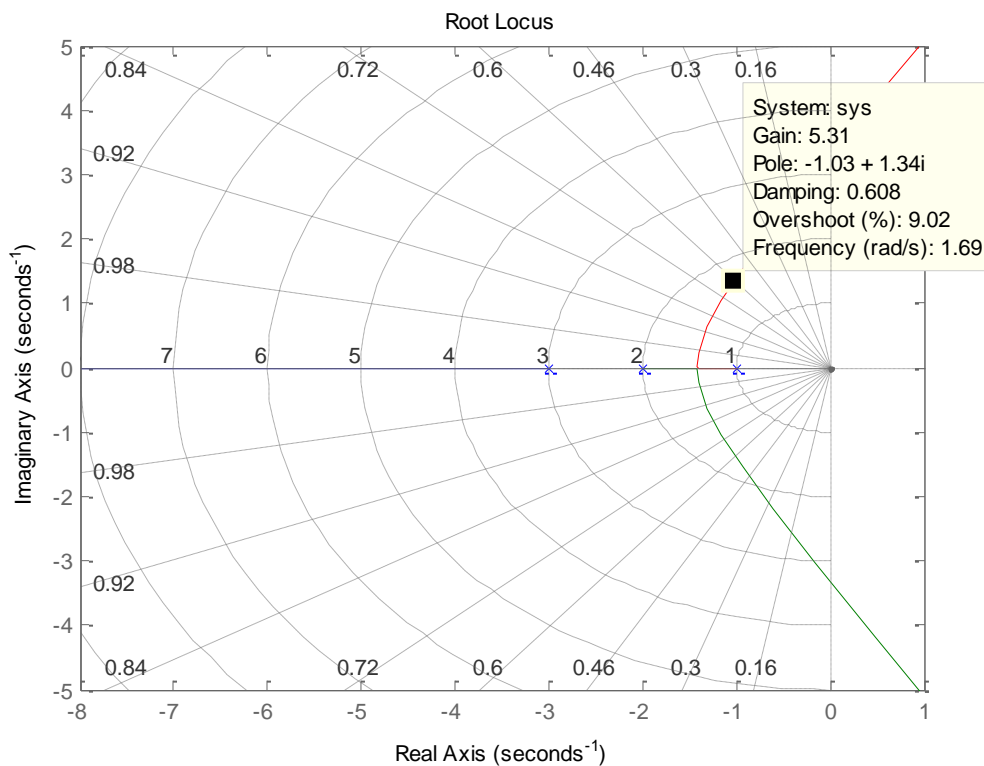
$$\text{angle condition} = \sum_{n=1}^{n=3} [\theta_{zeros} - \theta_{poles}] = \pm 180(2R + 1) = 90 + 54 + 36 = 180 \text{ deg}$$

$$\text{magnitude condition} = \sum_{n=1}^{n=3} \frac{\|poles\|}{\|zeros\|} = K = 1.4 * 1.6 * 2.4 = 5.3$$

$$\sum_{n=1}^{n=3} \text{open loop poles} = \sum_{n=1}^{n=3} \text{closed loop poles} = \text{constant as } n - m \geq 2$$

$$\sum_{n=1}^{n=3} \text{open loop poles} = -1 - 2 - 3 = -6 = \sum_{n=1}^{n=3} \text{closed loop poles} = 2(-1) \pm j1.4 + p$$

then  $p = -4$  i. e. closed loop poles are  $[-1 \pm j1.4, -4]$



### 19- To find analytically closed loop poles and K as

$(S^2+2\zeta\omega_n S+\omega_n^2)(S+a)$ =characteristic equa. for a third order syst.

$$\begin{aligned} \text{Solve } 1+G(S)H(S)=0 &= S^3+6S^2+11s+6+K=(S^2+1.2\omega_n S+\omega_n^2)(S+a) \\ &= S^3+(1.2\omega_n+a)S^2+(1.2\omega_n a+\omega_n^2)S+\omega_n^2 a \end{aligned}$$

$$1.2\omega_n+a=6 \quad , , , \quad 1.2\omega_n a+\omega_n^2=11 \quad , , , \quad \omega_n^2 a=k+6$$

Prog. `>>n=[1];d=[1 6 11 6]; rlocus(n,d), grid`

Q3

(30marks)

Consider a control system shown in Fig.1 if  $H(s)=1/(S+2)$ ,  $G(s)=k/(S+1)(s+3)$ ,

- Prove that the gain margin=6.02db at 3.32rad/sec. and the phase margin=25.4 degrees at 2.35rad/sec. **as K=30?**
- Sketch the **polar plot as K=30?**
- Sketch the **Bode plot as K=30?**
- Find the gain margin and the phase margin using **the plots?**
- Write a short MATLAB program to solve b ,c and d?
- Write short MATLAB program to Sketch the **Nichols plot?**

1- the open loop TF= $G(s) H(s)= G(S) H(S)$   
 $=K/[(S+3)(S+2)(S+1)]=30/[S^3+6S^2+11s+6]$

2- Find the freq.open loop TF=

$$G(j\omega)H(j\omega)=\frac{30}{[(j\omega+3)(j\omega+2)(j\omega+1)]} = Me^{j\Phi} = M \angle \Phi = Re+j \text{ imag}$$

$$M = \frac{30}{\sqrt{1+\omega^2}\sqrt{4+\omega^2}\sqrt{9+\omega^2}} \quad ,$$

$$\Phi = -\tan^{-1}(\omega) - \tan^{-1}(\omega/2) - \tan^{-1}(\omega/3)]$$

$$M = \frac{30}{\sqrt{1+\omega^2}\sqrt{4+\omega^2}\sqrt{9+\omega^2}} = \frac{30}{\sqrt{1+2.35^2}\sqrt{4+2.35^2}\sqrt{9+2.35^2}} = 1$$

$$M = \frac{30}{\sqrt{1+\omega^2}\sqrt{4+\omega^2}\sqrt{9+\omega^2}} = \frac{30}{\sqrt{1+3.32^2}\sqrt{4+3.32^2}\sqrt{9+3.32^2}} = 1$$

$$M = \frac{30}{\sqrt{1+\omega^2}\sqrt{4+\omega^2}\sqrt{9+\omega^2}} = \frac{30}{\sqrt{1+3.32^2}\sqrt{4+3.32^2}\sqrt{9+3.32^2}} = 0.5$$

$$\Phi = -\tan^{-1}(3.32) - \tan^{-1}(3.32/2) - \tan^{-1}(3.32/3)] = -180 \text{ deg.}$$

$$\Phi = -\tan^{-1}(2.35) - \tan^{-1}(2.35/2) - \tan^{-1}(2.35/3) = -154.6 \text{ deg.}$$

$$\gamma_m = \angle G(j\omega_g)H(j\omega_g) + 180 \text{ deg.} = 180 - 154.6 = 25.4 \text{ deg.}$$

3- Find the table

$\omega$	0	0.1	1	2.35	3.32	5	10	$\infty$
$\Phi$	0	-10.5	-90	-155	-180	-206	-236	-270
M	5	4.97	3	1	0.5	0.2	0.03	0
$20\log M$	14	13.9	9.5	0	6.4	-14.6	-31.1	0
Real $G(j\omega)H(j\omega)$	5		0					0
Imag $G(j\omega)H(j\omega)$	0		-3		0			0

4- Plot the vector on the  $j\omega$  – **plane** where  $\Phi$  in degrees as a straight line and determine M on this line

5- Plot the locus of the vector as points from the table

6- Find the gain and the phase margins from the plot

**Prog.** `>>n=[1]; d=[1 6 11 6];`

`>> nyquist(n,d)    >> margin(n,d)    >> nichols(n,d)`

