



جامعة بنها - كلية الهندسة ببناها - قسم الهندسة الكهربائية  
 نموذج الإجابة امتحان مادة هندسة التحكم ك352 تخلفات مايو السبب 2016-5-22  
 مدرس بالقسم شوقي حامد عرفه

Benha University, Faculty of Engineering Subject: Control Engineering (E352)	Elect.Eng.Depart. Third Year 22-5-2016 Time: 3-hours تخلفات  
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Q1 (10 mark)

- a- Define: gain margin- phase margin?
- b- Define:  $\omega_n$  ,  $\omega_d$  ,  $\omega_r$  ,  $\omega_c$  ,  $\omega_g$  ,  $\omega_p$  ,  $M_r$  ,  $\eta$ ?
- c- Consider a control system shown in Fig.1 if  $G(S) = K/[S^2+8S+7]$ ,  $H(S) = 1/(S+2)$ . Find the **open loop** transfer function and the **closed loop** transfer function?

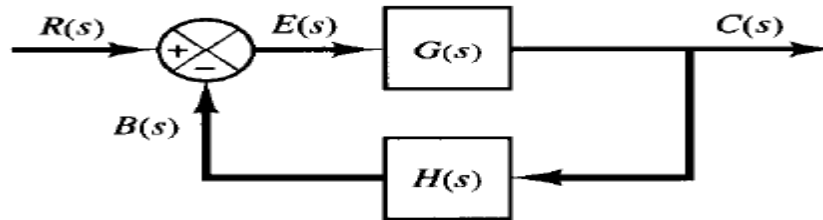


Fig.1

Q2 (15 marks)

Consider a unity feedback control system shown in Fig.1 if  $G(s) = \frac{16}{s(s+4)}$

- a-Find  $M_r$  and  $\omega_r$  ?
- b- Find the frequency response as  $r(t)=3\sin\omega t$  ?

Q3 (15 marks)

Consider a control system shown in Fig.1 if

$$G(s)H(s) = \frac{k}{(s+1+j)(s+2)(s+1-j)} = \frac{k}{s^3+4s^2+6s+4}$$

- a- Sketch the **complete root locus** for positive values of **K**?

- b- Find **K** that makes the complex closed loop poles have a damping ratio =**0.5** and **find the closed loop poles using the plot?**
- c- Find **K** that makes the complex closed loop poles have a damping ratio =**0.5** and **find the closed loop poles analytically?**
- d- Write short MATLAB program to solve a & b?

Q4

(20 marks)

Consider a control system shown in Fig.1 if

$$G(s)H(s) = \frac{10}{(S + 1 + j)(S + 2)(S + 1 - j)} = \frac{10}{S^3 + 4S^2 + 6S + 4}$$

- a- Prove that the gain margin=**6.02 db at 2.45 rad/sec.** and the phase margin=**30.3 degrees at 1.78 rad/sec.?**
- b- Sketch the **polar plot?**
- c- Sketch the **Bode plot?**
- d- Show the gain margin and the phase margin on **the plots?**
- e- Write short MATLAB program to solve a , b and C?

#### Answer

Q1

(10 marks)

a-Define: gain margin- phase margin?

**-Gain margin  $G_m$ :** it is reciprocal of the magnitude of the output frequency response at the Phase crossover frequency  $\omega_p$

$$a- G_m = 1/[\text{Real of } G(j \omega_p)H(j \omega_p)] = 1/|G(j \omega_p)H(j \omega_p)| = K_c/K$$

$$G_M = 20 \log G_m \text{ db}$$

**Phase margin  $\gamma_m$ :** it is the angle of the output frequency response at the -b gain crossover frequency plus 180 degrees -c

$$\gamma_m = \angle G(j \omega_g)H(j \omega_g) + 180 \text{ deg.}$$

b- Define:  $\omega_n$  ,  $\omega_d$  ,  $\omega_r$  ,  $\omega_c$  ,  $\omega_g$  ,  $\omega_p$  ,  $M_r$  ,  $\eta$ ? -d

**-Natural frequency  $\omega_n$  rad/sec:** it is the natural frequency depends on the natural of the system parameters.

**- Under damped natural frequency  $\omega_d$  rad/sec:** it is the under damped natural frequency depends on the damping coefficient  $\eta$  as it is less than one  $\eta < 1$ .

$$\omega_d = \omega_n \sqrt{1 - \zeta^2}$$

**-Resonant frequency  $\omega_r$  rad/sec:** it is the frequency at which the peak value of the output frequency response for a second order is occurred and it is equal to

$$\omega_r = \omega_n \sqrt{1 - 2\zeta^2}, \quad 0 \leq \zeta \leq 0.707$$

As  $\zeta$  approaches zero,  $M_r$  approaches infinity

$0 < \zeta \leq 0.707$ , the resonant frequency  $\omega_r$  is less than the damped natural frequency

**-Corner frequency  $\omega_c$  rad/sec:** it is the frequency at which the magnitude of the output frequency response is changed sharply. It may be  $(0, 1, 1/T, \omega_n)$

**-Gain crossover frequency  $\omega_g$ :** it is the frequency at which the magnitude of the output frequency response is equal to one or zero decibel.

$$|G(j\omega_g)H(j\omega_g)| = 1 \quad \text{or} \quad |G(j\omega_g)H(j\omega_g)| = 0 \text{ db}$$

**-Phase crossover frequency  $\omega_p$ :** it is the frequency at which the phase of the output frequency response is equal to  $(-180)$  degrees.

$$\text{Imag. } [G(j\omega_p)H(j\omega_p)] = 0 \quad \text{or} \quad \angle G(j\omega_p)H(j\omega_p) = -180 \text{ deg.}$$

**-Maximum resonant magnitude  $M_r$ :** it is the peak value of the output frequency response for a second order system  $M_r = \frac{1}{2\zeta \sqrt{1 - \zeta^2}}$

$$\frac{C(s)}{R(s)} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

$$M_r = |G(j\omega)|_{\max} = |G(j\omega_r)| = \frac{1}{2\zeta \sqrt{1 - \zeta^2}}$$

**-damping coefficient  $\eta$**  it depends on the natural of the system parameters. For second order system

$$\frac{C(s)}{R(s)} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

Values of $\eta$	System stability	Step-response
$0 > \eta$	System is unstable	undefined
$\eta = 0$	System is critically stable	oscillatory
$0 < \eta < 1$	System is stable	Under-damped
$0 < \eta = 1$	System is stable	Critically damped

$0 < \eta < 1$	System is stable	Over damped
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c- Consider a control system shown in Fig.1 if  $G(S) = K/[(S^2+8S+7)]$ ,  $H(S) = 1/(S+2)$ . Find the **open loop** transfer function and the **closed loop** transfer function?

The open loop TF= $G(s) H(s) = K/[(S^2+8S+7)(S+2)]$

$$\text{Closed loop tf} = \frac{C(S)}{R(S)} = \frac{G(s)}{1 + G(S)H(S)} = \frac{K(S+2)}{S^3 + 10S^2 + 23S + 14 + K}$$

Q2 (15 marks)

Consider a unity feedback control system shown in Fig.1 if  $G(s) = \frac{16}{s(s+4)}$

a-Find  $M_r$  and  $\omega_r$  ?

$$\frac{C(S)}{R(S)} = \frac{G(s)}{1 + G(S)H(S)} = \frac{\omega_n^2}{S^2 + 2\eta\omega_n S + \omega_n^2} = \frac{16}{S^2 + 4S + 16}$$

$$\omega_n = 4 \text{ rad/sec.}, \eta = 0.5, M_r = \frac{1}{2\zeta \sqrt{1 - \zeta^2}} = \frac{1}{2(0.5) \sqrt{1 - (0.5)^2}} = 1.155$$

$$\omega_r = \omega_n \sqrt{1 - 2\zeta^2} = 4\sqrt{1 - 2(0.5)^2} = 2.818 \text{ rad/sec.}$$

b- Find the frequency response as  $r(t) = 3\sin\omega t$  ?

Steps to find frequency Response:

1- the closed loop transfer function =  $T(s) = C(S)/R(S) =$

$$\frac{C(S)}{R(S)} = \frac{G(s)}{1 + G(S)H(S)} = \frac{\omega_n^2}{S^2 + 2\eta\omega_n S + \omega_n^2} = \frac{16}{S^2 + 4S + 16}$$

2-the closed loop frequency transfer function =

$$T(j\omega) = C(j\omega)/R(j\omega) = \frac{16}{(j\omega)^2 + 4(j\omega) + 16} = M \angle \Phi = \text{Re} + j \text{imag}$$

$$M = \frac{16}{\sqrt{(16 - \omega^2)^2 + 16\omega^2}}, \quad \Phi = \tan^{-1}[4\omega / (16 - \omega^2)]$$

3-As the input =  $r(t) = 3\sin\omega t$  then

$$\begin{aligned} \text{the response} &= C(t) = 3M \sin(\omega t + \Phi) \\ &= \frac{48}{\sqrt{(16 - \omega^2)^2 + 16\omega^2}} \sin[\omega t + \tan^{-1}[4\omega / (16 - \omega^2)]] \end{aligned}$$

Q3

(15 marks)

Consider a control system shown in Fig.1 if

$$G(s)H(s) = \frac{k}{(s+1-j)(s+2)(s+1+j)} = \frac{k}{s^3+4s^2+6s+4}$$

- Sketch the **complete root locus** for positive values of **K**?
- Find **K** that makes the complex closed loop poles have a damping ratio =**0.5** and **find the closed loop poles using the plot**?
- Find **K** that makes the complex closed loop poles have a damping ratio =**0.5** and **find the closed loop poles analytically**?
- Write short MATLAB program to solve a & b?

a-Root locus:

1-the root locus is symmetrical about the real axis in the S-plane

$$2\text{-the open loop } G(s)H(s) = \frac{k}{(s+1-j)(s+2)(s+1+j)} = \frac{k}{s^3+4s^2+6s+4}$$

3-the root locus starts at the pole and ends at the zeros or infinity

4-number of root loci= n=number of poles of the open loop TF =3 at [-1+j,-1-j,-2]

5-number of zeros= m=0

6-number of asymptotes = n-m=3-0=3

8-center of gravity =  $A = \frac{\sum \text{poles} - \sum \text{zeros}}{n-m} = \frac{-1-1-2}{3} = -1.3$  point of intersection of asymptotes with real axis=

9-angles of asymptotes are =  $\Theta = \frac{\pm 180(2R+1)}{n-m} = \pm 60, \pm 180$

10- Points of crossing the imaginary axis as Routh test

$$\text{Charct. equa} = 1 + G(S)H(S) = 0 = s^3 + 4s^2 + 6s + 4 + K$$

$S^3$	1	6	$4+K \geq 0, [20-K]/4 \geq 0$ then $-4 \leq K \leq 20, K_c = 20$ $4S^2 + 24 = 0, S = \pm j \omega = \pm \sqrt{6}$ rad/sec
$S^2$	4	$4+K$	
$S$	$[24-4-K]/4$		
$S^0$	$4+K$		

11- there is no break points (break away or break in) at

$$-\frac{dK}{dS} = 0 = \frac{d}{dS} \left[ \frac{1}{G(S)H(S)} \right] = \frac{d}{dS} [S^3 + 4S^2 + 6s + 4] = 3S^2 + 8s + 6 = 0$$

12- There is no break angles  $[\pm 180(2R+1)/r]$  where  $r$ =number of branches (poles for break away or zeros for break in)  $R=0,1,-----$ no break angles

13-the angle of departures (complex poles) =

**Angle of departure from a complex pole –  $180^\circ$**

– (sum of the angles of vectors to a complex pole in question from other poles)

+ (sum of the angles of vectors to a complex pole in question from zeros)

angle of departure =  $\pm 180 - 90 - 45 = \pm 45 \text{ deg}$

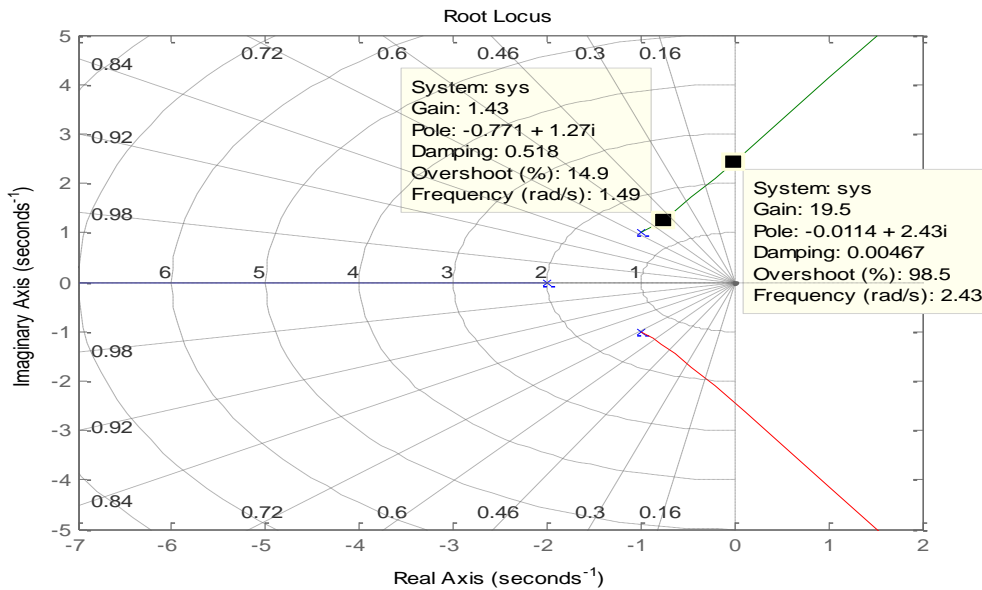
14-angle of arrival (complex zeros) as

**Angle of arrival at a complex zero =  $180^\circ$**

– (sum of the angles of vectors to a complex zero in question from other zeros)

+ (sum of the angles of vectors to a complex zero in question from poles)

15-sketch the root loci as



16- the damping factor or coefficient  $\zeta$  is straight line with slope  $\Theta = \cos^{-1}\zeta$

with respect to the negative real axis in the S-plane.  $\Theta = \cos^{-1} 0.5 = 60\text{deg}$ . at the test point (intersection point)  $S_d = -0.8 \pm j1.3$

$$\text{angle condition} = \sum_{n=1}^{n=3} [\theta_{zeros} - \theta_{poles}] = \pm 180(2R + 1) = 90 + 54 + 36 = 180 \text{ deg}$$

$$\text{magnitude condition} = \sum_{n=1}^{n=3} \frac{\|poles\|}{\|zeros\|} = K = * * = 1.5$$

$$\sum_{n=1}^{n=3} \text{open loop poles} = \sum_{n=1}^{n=3} \text{closed loop poles} = \text{constant as } n - m \geq 2$$

$$\begin{aligned} \sum_{n=1}^{n=3} \text{open loop poles} &= -1 - 2 - 1 = -4 = \sum_{n=1}^{n=3} \text{closed loop poles} \\ &= (-0.8 + j1.3, -0.8 - j1.3, p) \end{aligned}$$

then  $p = -2.4$  i. e. closed loop poles are  $[-0.8 \pm j1.3, -2.4]$

**19- To find analytically closed loop poles and K as**

$(S^2 + 2\zeta\omega_n S + \omega_n^2)(S+a) = \text{characteristic equa. for a third order syst.}$

Solve  $1+G(S)H(S)=0 = S^3+4S^2+6s+4+K=(S^2 + \omega_n S + \omega_n^2)(S+a)$   
 $= S^3+(\omega_n+a)S^2+(\omega_n a + \omega_n^2)S + \omega_n^2 a$

$\omega_n + a = 4$  ,  $\omega_n a + \omega_n^2 = 6$  ,  $\omega_n^2 a = k+4$ , then  $\omega_n = 1.5 \text{ rad/sec.}$ ,  $a = 2.5$ ,  $k = 1.74$

Prog. `>>n=[1];d=[1 4 6 4]; rlocus(n,d), grid`

Q4 (20 marks)

Consider a control system shown in Fig.1 if

$$G(s)H(s) = \frac{10}{(S + 1 + j)(S + 2)(S + 1 - j)} = \frac{10}{S^3 + 4S^2 + 6S + 4}$$

- a- Prove that the gain margin=**6.02 db at 2.45 rad/sec.** and the phase margin=**30.3 degrees at 1.78 rad/sec.?**
  - b- Sketch the **polar plot?**
  - c- Sketch the **Bode plot?**
  - d- Show the gain margin and the phase margin on **the plots?**
  - e- Write short MATLAB program to solve a , b and C?
- 1- the open loop TF= $G(s) H(s) = \mathbf{G(S) H(S)}$

$$G(s)H(s) = \frac{10}{(S + 1 + j)(S + 2)(S + 1 - j)} = \frac{10}{S^3 + 4S^2 + 6S + 4}$$

2- Find the freq.open loop TF=

$$\mathbf{G(j\omega)H(j\omega)} = \frac{10}{S^3+4S^2+6S+4} = Me^{j\Phi} = M \angle \Phi = \text{Re} + j \text{imag}$$

$$M = \frac{10}{\sqrt{(4-4\omega^2)^2 + (6\omega-\omega^3)^2}} , \Phi = -\tan^{-1}((6\omega - \omega^3)/(4 - 4\omega^2))$$

$$M = \frac{10}{\sqrt{(4-4\omega^2)^2 + (6\omega-\omega^3)^2}} = \frac{10}{\sqrt{(4-4\omega^2)^2 + (6\omega-\omega^3)^2}} = 1$$

$$M = \frac{10}{\sqrt{(4-4(1.78)^2)^2 + (6(1.78)-(1.78)^3)^2}} = 1, \text{ then } \omega_g = 1.78 \text{ rad/sec.}$$

$$\Phi = -\tan^{-1}((6\omega - \omega^3)/(4 - 4\omega^2)) = -\tan^{-1}((6 * 2.45 - 2.45^3)/(4 - 4 * 2.45^2)) = -180 \text{ deg.}$$

then  $\omega_p = 2.45 \text{ rad/sec.}$



$$M = \frac{10}{\sqrt{(4 - 4(2.45)^2)^2 + (6(2.45) - (2.45)^3)^2}} = 0.5, \text{ then } G_M = 20 \log \frac{1}{0.5} = 6.02 \text{ db}$$

$$\Phi = -\tan^{-1}((6\omega - \omega^3)/(4 - 4\omega^2)) = -\tan^{-1}((6 * 1.78 - 1.78^3)/(4 - 4 * 1.78^2)) = -149.7 \text{ deg.}$$

$$\gamma_m = \angle G(j\omega_g)H(j\omega_g) + 180 \text{ deg.} = 180 - 149.7 = 30.3 \text{ deg.}$$

3- Find the table

$\omega$	0	0.1	1	1.78	2.45	5	10	$\infty$
$\Phi$	0			-150	-180	-		-270
M	2.5		2	1	0.5			0
20logM	8		-6.02	0	6.02	-	-	0
Real $G(j\omega)H(j\omega)$	2.5		0		-0.5			0
Imag $G(j\omega)H(j\omega)$	0		-2		0			0

4- Plot the vector on the  $j\omega - \text{plane}$  where  $\Phi$  in degrees as a straight line and determine M on this line

5- Plot the locus of the vector as points from the table

6- Find the gain and the phase margins from the plot

**Prog.** >>n=[10]; d=[1 4 6 4];

>> nyquist(n,d) >> margin(n,d) >> nichols(n,d)

